

Multiple Antennas

—

Capacity and Basic Transmission Schemes



Mats Bengtsson, Björn Ottersten

Basic Transmission Schemes

1

September 8, 2005

PRESENTATION OUTLINE

- Channel capacity — Some fine details and misconceptions
- Basic transmission schemes
 - Single link without CSI@Tx
 - Single link with CSI@Tx
 - Multiple users
- Later talks (not included here!):
 - Space-Time block coding
 - Exploiting partial channel information

(CSI = “Channel State Information”)



Basic Transmission Schemes

2

September 8, 2005

CHANNEL CAPACITY

Definition: The highest data rate where the bit error rate can $\rightarrow 0$ when the code word length $\rightarrow \infty$.



Theorem: Under certain conditions, channel capacity = maximum mutual information between transmitter and receiver.

Theorem: Under certain conditions, the mutual information is maximized when the code words are Gaussian distributed.

Note: Channel statistics are always assumed known (also at the transmitter).

MIMO CAPACITY, AWGN

Channel assumption: $\mathbf{y} = \mathbf{H}\mathbf{x} + \mathbf{n}$,
 \mathbf{H} deterministic fixed, $\mathbf{n} \in \mathcal{CN}(0, \mathbf{R}_n)$.

Power constraint: $E[\|\mathbf{x}\|^2] = \text{Tr}[\mathbf{R}_x] \leq P_{\max}$

Capacity:



$$\begin{aligned} C &= \max_{\text{Tr}[\mathbf{R}_x] \leq P_{\max}} \log \det |\mathbf{R}_n + \mathbf{H}\mathbf{R}_x\mathbf{H}^*| - \log \det |\mathbf{R}_n| \\ &= \max_{\text{Tr}[\mathbf{R}_x] \leq P_{\max}} \log \det |\mathbf{I} + \mathbf{H}\mathbf{R}_x\mathbf{H}^*\mathbf{R}_n^{-1}| \end{aligned}$$

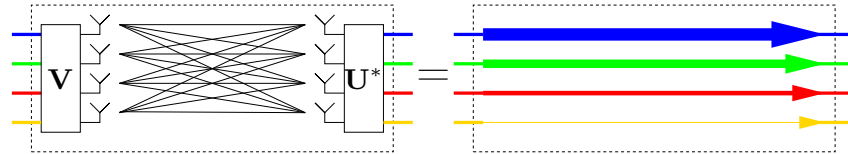
Note: Maximum easily found using “water filling”

Note: It does not make sense to talk about “capacity of an AWGN with no CSI@Tx”!

MIMO CAPACITY, AWGN, INTERPRETATION

Assume: $\mathbf{R}_n = \sigma_n^2 \mathbf{I}$

Singular Value Decomposition: $\mathbf{H} = \mathbf{U}\mathbf{S}\mathbf{V}^*$



Equivalent channel: $\tilde{\mathbf{y}} = \mathbf{U}^* \mathbf{y} = \mathbf{S} \underbrace{\mathbf{V} \mathbf{x}}_{\tilde{\mathbf{x}}} + \underbrace{\mathbf{V} \mathbf{n}}_{\tilde{\mathbf{n}}} = \mathbf{S} \tilde{\mathbf{x}} + \tilde{\mathbf{n}} \iff$
 $\tilde{y}_k = \sigma_k \tilde{x}_k + \tilde{n}_k, k = 1, \dots, \min\{N_{Tx}, N_{Rx}\}.$

Allocate power and rate over orthogonal scalar channels!

MIMO ERGODIC CAPACITY

Definition: The **ergodic capacity** is the channel capacity if

- Channel is fast fading, ergodic stochastic process.
- $M \rightarrow \infty$ channel realizations in each transmitted code word.

Note: The transmitted code words are designed only based on the statistics of the channel!



Theorem:

$$\begin{aligned} C &= \max_{\text{Tr}[\mathbf{R}_x] \leq P_{\max}} \mathbb{E} \{ \log \det |\mathbf{R}_n + \mathbf{H} \mathbf{R}_x \mathbf{H}^*| - \log \det |\mathbf{R}_n| \} \\ &= \max_{\text{Tr}[\mathbf{R}_x] \leq P_{\max}} \mathbb{E} \{ \log \det |\mathbf{I} + \mathbf{H} \mathbf{R}_x \mathbf{H}^* \mathbf{R}_n^{-1}| \} \end{aligned}$$

Theorem: If \mathbf{H} has i.i.d Rayleigh elements and $\mathbf{R}_n = \sigma_n^2 \mathbf{I}$, then

$$C = \mathbb{E} \left\{ \log \det \left| \mathbf{I} + \frac{P_{\max}}{\sigma_n^2 N_{Tx}} \mathbf{H} \mathbf{H}^* \right| \right\}$$

MIMO OUTAGE CAPACITY

Scenario:

- Block fading. Length of each fading block $\rightarrow \infty$.
- The transmitted code words are designed optimally for a fixed rate, only based on the statistics of the channel!

Definition: The outage probability is the probability that transmission within a fading block can be done with bit error rate $\rightarrow 0$.



Theorem: The outage probability at rate R is given by

$$p_{\text{out}}(R) = \min_{\text{Tr}[\mathbf{R}_x] \leq P_{\text{max}}} P \left\{ \log \det |\mathbf{I} + \mathbf{H}\mathbf{R}_x\mathbf{H}^*\mathbf{R}_n^{-1}| < R \right\}$$

Theorem: If \mathbf{H} has i.i.d Rayleigh elements and $\mathbf{R}_n = \sigma_n^2 \mathbf{I}$, the outage probability at rate R is given by

$$p_{\text{out}}(R) = P \left\{ \log \det \left| \mathbf{I} + \frac{P_{\text{max}}}{\sigma_n^2 N_{Tx}} \mathbf{H}\mathbf{H}^* \right| < R \right\}$$

MIMO SINGLE LINK – BASIC OPTIONS

- Increased data rate – Spatial Multiplexing
- Increased robustness to fading – Diversity

Trade-off Multiplexing \leftrightarrow diversity



Asymptotic multiplexing gain r : $R = r \log \text{SNR}$, when $\text{SNR} \rightarrow \infty$.

Diversity gain d : $p_{\text{out}}(\text{SNR}) \approx \text{SNR}^{-d}$, when $\text{SNR} \rightarrow \infty$.

Max multiplexing gain: $r_{\text{max}} = \min\{N_{Tx}, N_{Rx}\}$

Max Diversity gain: $d_{\text{max}} = N_{Tx}N_{Rx}$

Fundamental trade-off: Optimal diversity gain at multiplexing gain r : $d(r) = (N_{Tx} - r)(N_{Rx} - r)$

MIMO SINGLE LINK – NO CSI@Tx



Multiplexing Schemes: Transmit different data streams on different “antennas”. Use multi-user receive techniques to separate them. Requires non-linear processing. V-BLAST, D-BLAST, ...

Diversity Schemes: Introduce redundancy over space and time. Alamouti, Space-Time block codes, Space-time trellis codes, ...

(**Note:** Additional structure from diversity schemes may also be used for interference suppression.)

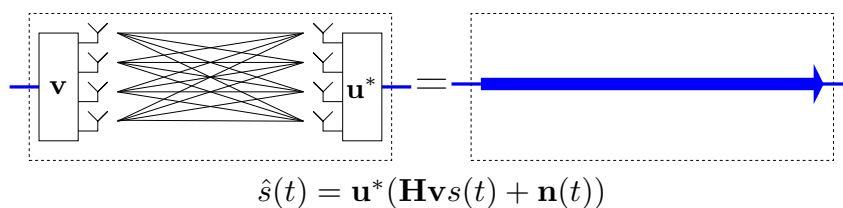
MIMO SINGLE LINK – CSI@Tx



Multiplexing Schemes: Create spatial subchannels, transmit different data streams over the subchannels. Only linear processing required.

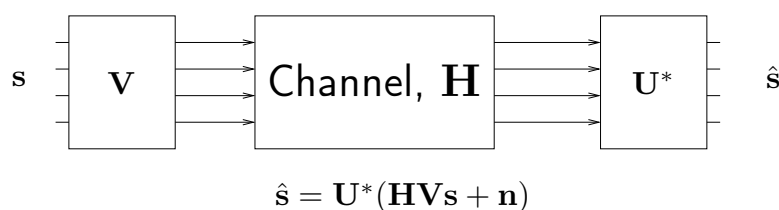
Diversity: Obtained by adapting the transmitted signal to the channel. Beamforming, ...

SINGLE LINK CSI@Tx, BEAMFORMING



- Single data stream!
- Optimal beamformers: left and right hand singular vectors with highest singular value.
- Diversity order $N_{Tx}N_{Rx}$ for i.i.d. Rayleigh fading channel.
- Note: Full channel state information needed at the transmitter (or feedback of Tx beamformer \mathbf{v}).
- Note: Same diversity order can be obtained without any CSI@Tx!

SINGLE LINK CSI@Tx, LINEAR PROCESSING



\mathbf{U} : Linear transmit processing

\mathbf{V} : Linear receive processing

Strategies for \mathbf{U} and \mathbf{V} :

Zero-forcing: $\mathbf{U}^*\mathbf{H}\mathbf{V} = \mathbf{I} \implies \hat{s} = \mathbf{s} + \mathbf{U}^*\mathbf{n}$. Risk for noise amplification!

MMSE: $\min_{\mathbf{U}, \mathbf{V}} E\|\hat{s} - \mathbf{s}\|^2$

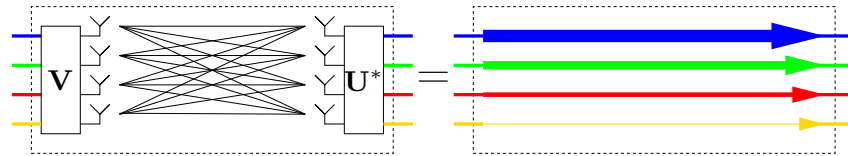
Minimum Bit Error Rate: ...

...

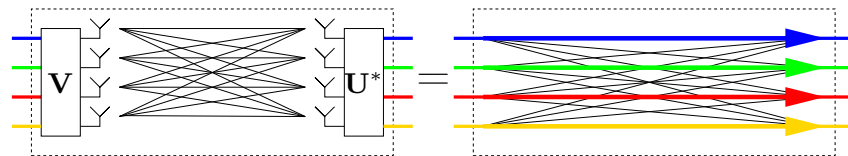
SINGLE LINK CSI@Tx, UNIFIED ANALYSIS

Two main transmit solutions!

Shur-Concave Criteria: MMSE, Max mutual information, mean SINR across subchannels, ...



Shur-Convex Criteria: Max MSE across subchannels, mean&max BER across subchannels, ...



MMSE Receiver is always optimum.

RECEIVER STRUCTURES IN GENERAL Maximum Likelihood

Approximate Maximum Likelihood, examples:

- Sphere decoding
- Semidefinite relaxations
- Lattice-Reduction-Aided Detectors



Other non-linear techniques, examples:

- Successive interference cancellation
- Parallel interference cancellation
- Iterative “Turbo decoding” techniques

Linear techniques, examples:

- MMSE
- Zero-Forcing
- Maximum Ratio Combining (conventional beamforming)

DIFFERENT POWER CONSTRAINTS

Total Employed Transmit Power: Most common choice!

$$P_{\text{tot}} = \text{Tr}[\mathbf{V}\mathbf{V}^*] \leq P_{\text{max}}$$

Maximum Element Power:

$$P_{\text{element}} = \max_{l=1,\dots,m} \|\mathbf{V}_{l,:}\|^2 \leq P_{\text{max}}$$



Equivalent isotropic radiated power (EIRP):

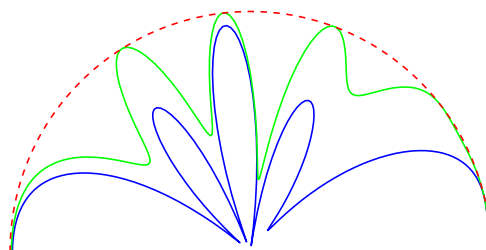
$$EIRP = \max_{\theta} \mathbf{a}(\theta)^* \mathbf{V}\mathbf{V}^* \mathbf{a}(\theta) \leq P_{\text{max}}$$

(where $\mathbf{a}(\theta)$ is array response vector in free space)

Note: Different constraints \implies different transmit schemes are optimal.

EXAMPLE, EIRP BEAMFORMING, MIMO

- Use beamforming at transmitter and receiver \implies one spatial channel per MIMO link.
- Constraints on Equivalent Isotropic Radiated Power (EIRP) at the transmitter.
- Maximize the received power.



Beam pattern of:

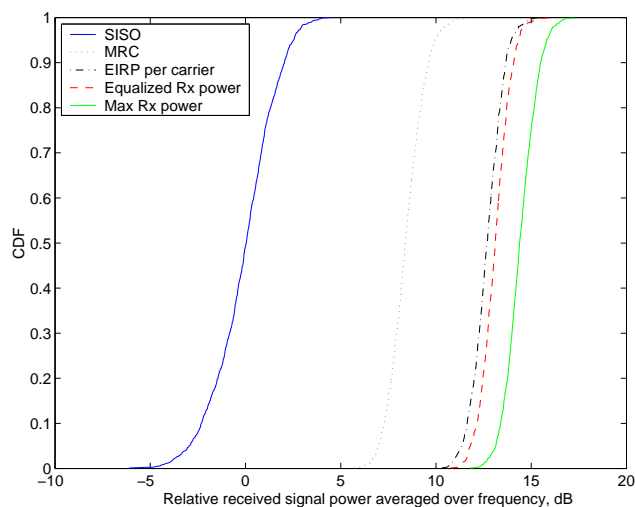
- Conventional BF
- Optimal EIRP BF
- - - EIRP limit

EIRP BEAMFORMING, PROPERTIES

- Fixed transmit beamformer \Rightarrow standard MRC receive beamformer optimum.
- Fixed receive beamformer \Rightarrow optimal transmit beamformer found using convex optimization.
- Trick: iterate!
- Easily generalized to OFDM.
- Can provide antenna gain of 10-15dB compared to single antenna transmitter.



EIRP BEAMFORMING, EXAMPLE MIMO+OFDM



Multicarrier 4×4 MIMO beamforming with EIRP constraints.

EXPLOITING THE SPATIAL DIMENSION FOR SPECTRAL EFFICIENCY – SYSTEM CAPACITY



- Interference rejection capabilities
Tighter frequency reuse
- Multiple users sharing same resource (time, frequency, code)
Spatial division multiple access
- Link capacity can be increased
TX and RX diversity in MIMO systems

LINEAR RECEIVERS IN MULTI-USER SYSTEMS

General Strategy:

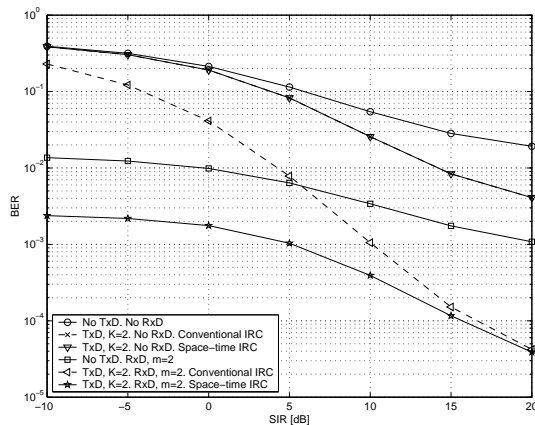
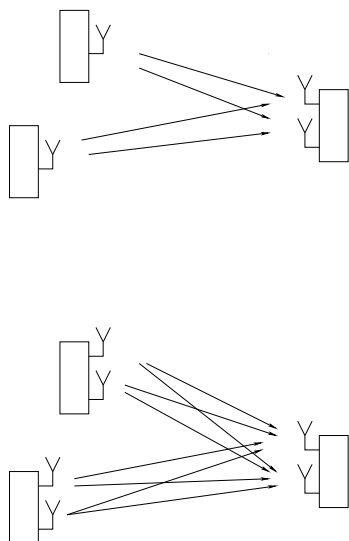
- Model interference plus noise as (spatially and/or temporally) colored noise. Characterized by:
 - Covariance matrix or
 - Vector valued AR-model or ...
- Use linear MMSE receiver.



Exploit structure in desired signal and/or interference!

Example: Alamouti coded interference has structure that can be exploited!

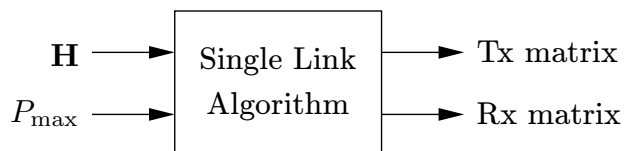
MIMO MULTI-USER ALAMOUTI EXAMPLE



Two users, with and without TX diversity and RX diversity.

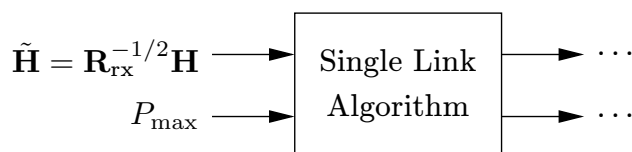
INTERFERENCE SUPPRESSION USING PREWHITENING

Assume given: Algorithms for a single link in white noise. E.g. optimize MMSE or SINR or mutual information or BER or ...



Interference suppression at the receivers:

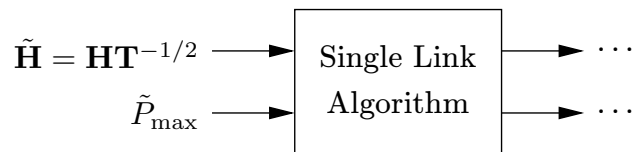
Noise prewhitening at the receivers (well-known).



INTERFERENCE SUPPRESSION, CONT.

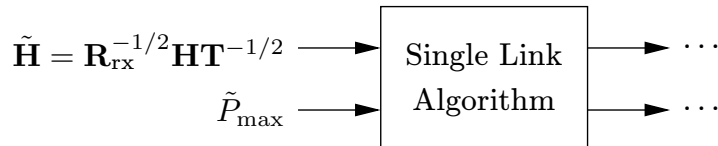
Interference suppression at the transmitters:

“Noise prewhitening” at the transmitters.



KTH Electrical Engineering

Combine:



Iterate! Circular dependence,

\mathbf{R}_{rx} and $\mathbf{T} \iff$ Tx and Rx parameters.

Solve iteratively for each link! Hope for convergence!

INTERFERENCE SUPPRESSION, CONT.

Choice of Transmit Preshwhitening Matrix \mathbf{T} ?

- Theoretical result: Some choice of \mathbf{T} gives optimal system performance.
- Practical proposal, for systems with linear Tx, Rx matrices:
 - Each receiver sends pilot data using its receiver matrix.
 - Each transmitter collects the pilot data and estimates the resulting covariance matrix. Use as \mathbf{T} !

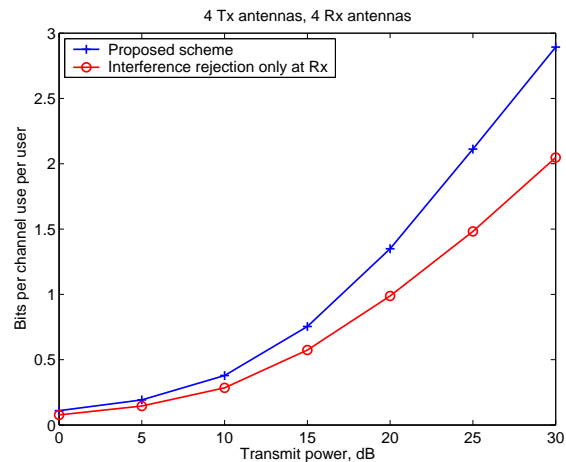


KTH Electrical Engineering

INTERFERENCE SUPPRESSION, EXAMPLE



- Max mutual information per link.
- Narrowband.
- 3GPP suburban channel model, $\pm 120^\circ$ sectors.
- 20 mobiles placed randomly in 7×3 cells.



WHAT INFORMATION THEORY SAYS ON MULTI-USER MIMO



- Very few results on general interference channel (multiple cells).
- Special cases where capacity region is known:
 - Multiple Access Channel “MAC” (single cell uplink)
 - Broadcast Channel “BC” (single cell downlink)
- Approaching these limits requires non-linear processing (“dirty-paper” coding, ...)

OBSERVATIONS FOR DUAL ARRAY STRUCTURES

- Capacity is not limited by frequency bandwidth.
- Multiple RX/TX at terminal feasible.
- High performance requires “rich scattering” environment.
- Systems must operate (and be designed for) operation with partial/limited channel knowledge.
- Efficient exploitation of limited channel knowledge essential for efficient designs, intricate problem for multi-user MIMO systems.
- Can trade complexity (power) between access point and terminal.



References

- [1] D. Tse and P. Viswanath. *Fundamentals of Wireless Communication*. Cambridge University Press, 2005.
- [2] D. P. Palomar et al. Joint Tx-Rx beamforming design for multicarrier MIMO channels: A unified framework for convex optimization. *IEEE Trans. SP*, 51(9):2381–2401, Sept. 2003.
- [3] P. Zetterberg et al. Downlink beamforming with delayed channel estimates under total power, element power and equivalent isotropic radiated power (EIRP) constraints. In *Proc. IEEE VTC Fall 2001*
- [4] G. Klang and B. Ottersten. Interference robustness aspects of space-time block code-based transmit diversity. *IEEE Trans. SP*, pages 1299–1309, Apr. 2005.
- [5] M. Bengtsson. A pragmatic approach to multi-user spatial multiplexing. In *Proc. IEEE SAM 2002*.

